PCI Express* 3.0 Technology: PHY Implementation Considerations for Intel Platforms

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Agenda

• Problem Statement
• Existing Usage of K-Code in 8b/10b
• Encoding Scheme
• Transmitter Equalization and training
• Implementation Considerations
• Summary

Disclaimer: Information contained herein is derived from Intel technology path finding and is Work In Progress and is subject to change
PCI Express* (PCIe*) Technology Roadmap

<table>
<thead>
<tr>
<th>Year</th>
<th>Raw Bit Rate</th>
<th>Link BW</th>
<th>BW/lane/way</th>
<th>BW x16</th>
</tr>
</thead>
<tbody>
<tr>
<td>1999</td>
<td>PCIe 1.x</td>
<td>2.5GT/s</td>
<td>2Gb/s</td>
<td>~250MB/s</td>
</tr>
<tr>
<td>2001</td>
<td>PCIe 2.0</td>
<td>5.0GT/s</td>
<td>4Gb/s</td>
<td>~500MB/s</td>
</tr>
<tr>
<td>2003</td>
<td>PCIe 3.0</td>
<td>8.0GT/s</td>
<td>8Gb/s</td>
<td>~1GB/s</td>
</tr>
</tbody>
</table>

Based on x16 PCIe channel

Continuous Improvement: Doubling Bandwidth & Improving Capabilities Every 3-4 Years!
Problem Statement

- **PCI Express* (PCIE*) 3.0 data rate decision: 8 GT/s**
  - High Volume Manufacturing channel for client/ servers
    - Same channels and length for backwards compatibility
    - Low power and ease of design
      - Avoid using complicated receiver equalization, etc.

- **Requirement: Double Bandwidth from Gen 2**
  - PCIE 1.0a data rate: 2.5 GT/s
  - PCIE 2.0 data rate: 5 GT/s
    - Doubled the data rate/bandwidth from Gen 1 to Gen 2
  - Data rate gives us a 60% boost in bandwidth
  - Rest will come from **Encoding**
    - Replace 8b/10b encoding with a scrambling-only encoding scheme when operating at PCIE 3.0 data rate

- **Double B/W: Encoding efficiency 1.25 X data rate 1.6 = 2X**

**Challenge:** New Encoding Scheme to cover 256 data plus 12 K-codes with 8 bits
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Review of K-Code Usage

• Each K-code is a unique 10-bit value
  – Distinct from data and other K-codes

• Two flavors for K-code use
  – Packet Stream (independent of link width)
  – Lane Stream (per-lane)

• Packet Stream relates to Packet Framing (Link-Wide)
  – STP - Start of Transaction Layer Packet (TLP)
  – END - End (Good) of TLP
  – EDB - End Bad of TLP
  – SDP - Start of Data Link Layer packet (DLLP)

• Lane Stream relates to Ordered Sets:
  – Training Set #1 & #2: Training/ retraining
  – SKP Ordered Sets: clock compensation and byte realignment
  – Electrical Idle Start/ Exit sequence: Power Management

• New encoding scheme accommodates these existing usages

Functionality of K-Code needs to be preserved
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Requirements and Capabilities

- **Basic Fault Model:**
  - Guaranteed error detection against random bit flips in any packet or Ordered Set
- **Eventual recovery from bit slip/add**
- **Handle killer packets**
  - Send a different bit stream on retry of a packet
- **Low bandwidth overhead (1-2%)**
- **Low L0s/L1 exit latency overhead**
  - Preserve aggressive power management with performance
- **Changes mostly limited to physical layer**
- **Protocol development concurrent with analysis / simulation done by Intel Pathfinding team**

New encoding scheme: Better Performance and Reliability than PCI Express* 2.0 Technology
LTSSM Speed Change: Example

LTSSM in Device A

L0 (2.5 G)
L0 -> Recovery (directed_speed_change)

Recovery.RcvrLock
TS1 w/ speed_change = 0
TS1 w/ speed_change = 1
(Gen 2 data rate advertised in TS1/TS2)

Recovery.RcvCfg
TS2 w/ speed_change = 1

Recovery.Speed

Recovery.LcvrLock
Link goes from Recovery.RcvrLock to L0 in 8 G

L0 (8 G)

Source: Intel Corporation

LTSSM in Device B

L0 (2.5 G)

Recovery.RcvrLock
Successful Speed Increase initiated by A

Recovery.RcvrLock
(directed_speed_change set after 8 TS1 rcvd)

Recovery.RcvCfg

Recovery.Speed

Recovery.LcvrLock

L0 (8 G)

Both Encoding Schemes Co-exist

8b/10b Encoding

Source: Intel Corporation
128b/130b Encoding Scheme

Two levels of encapsulation

- **Lane Level: Blocks**
  - Data vs Ordered Sets
  - 2-bit Sync Header identifies Data Block vs OS (not scrambled)
  - 128-bit payload
  - Rationale:
    - Redundancy helps separate Data from OS
    - 128-bit payload chosen to match OS payload. 2-bit is low overhead for Sync header is low

- **Data Block: Link wide with Framing preamble identifying packet boundary up-front**
  - Multiple packets within a Data Block and one packet can straddle multiple Blocks
  - Framing preamble same overhead as in 8b/10b
  - Payload scrambled

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Scrambling with two levels of encapsulation
Data Block

- Packets: Logical IDL (LIDL), DLLP, TLP, etc.
- Various sizes from 8b/10b time: 1 Symbol for LIDL, DLLP: 8 Symbols, TLP: Multiple of 4 Symbols
- Everything other than TLP is fixed size
- Need to ensure triple bit flip detection ability while keeping the sizes the same
  - New encoding: One Symbol becomes 8bit rather than 10bit in 8b/10b
- TLP and DLLP body is LCRC protected which provides triple bit flip detection ability
  - Framing preamble itself needs to ensure triple bit flip detection ability as it is used to determine packet boundary
- Challenge: Framing preamble itself has to be of variable length
- Solution: Use first Symbol encoding between different entities to be at a Hamming distance of 4 to ensure triple bit flip detection ability
  - Subsequent Symbols, if any, use some form of CRC protection itself
- Robustness features confirmed by analysis/simulations in Intel path finding
Example of TLP Transmission in a X4

(TLP Transmitted: 3 DW Header (h0 .. h11) + 1 DW Data (d0 .. D3).
1 DW LCRC (L0 .. L3) and Q[11:0]: Sequence No from Link Layer)

[Framer O/P: STP S[3:0] = f h; length l[10:0] = 006h;
Length CRC C[3:0] = f h; Parity P = 0b]
Ordered Sets

- Used for Link training, power management, clock compensation
- Additional usage: Block alignment
  - Can not be scrambled
  - Must reset scrambler (so that both sides start at same point)
  - Can not do this during Data Blocks
    - Unlike 8b/10b where COM can never be aliased to between any two 10-bit Symbols, one can easily alias to any bit pattern when two Data Blocks are looked at consecutively for a fixed pattern
    - Must be done when all permutations are not possible (i.e., during Recovery)
    - Choice of Ordered Set encoding to be such that one can always correctly do block alignment
    - Another challenge is bit slip initially, based on past observation
    - Solution: Continuously do block alignment in Recovery while Ordered Sets are on
- Ordered Set for Clock compensation
  - Can not be scrambled (e.g., repeaters)
  - Need to carry information such as LFSR value to help trace tools
- Link training needs a spectrally rich pattern for better bit lock
  - Solution: Most of TS1/TS2 are scrambled; Rest are not
- Problem scenarios and their solutions created and verified in analysis / simulation in Intel pathfinding
Sync Header Protection

• Challenge: Sync Header is 2-bits: so even a two bit flip can potentially change a Data Block to an Ordered Set and vice-versa. Even worse the LFSR can be out of sync between the Tx and the Rx and introduce massive errors as a result and cause data corruption

• Solution: Define a “marker” of the appropriate type whenever there is a transition from Data Block to Ordered Set and vice-versa
  - Pre-notification of the Block type change – that itself is adequately protected
  - Data Block to OS: In last Data Block as a CRC-protected packet
  - OS to Data Block: As an additional 130-bit marker OS
  - Protects more than triple bit flip
  - Problem scenario as well as solution created and validated through analysis/ simulation (Intel path finding)
Error Recovery

• Framing error detected by PHY
  – Helps identify cases where the physical layer can either have its scrambler out of sync or fails to ascertain the next packet’s framing preamble location

• Any framing error directs LTSSM to Recovery
  – Stop processing any received TLP/DLLP after Recovery to avoid data corruption
    • The CRC within these packets become ineffective when the packet boundary is lost – random data can always alias to a good CRC
  – Block lock and scrambler reset happens through Recovery prior to packet being accepted
  – Link layer detected errors can be recovered through packet retry

• Error Detection Guarantees maintained
  – Triple bit flip detection within each TLP/ DLLP/ IDL/ OS

Robust Error Detection and Recovery Mechanism required with PHY Framer
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Transmitter Equalization

- 2.5 GT/s: Same de-emphasis for all
- 5 GT/s: Introduced platform dependent de-emphasis selection on a per-Link basis
  - -3.5 dB and -6 dB
- 8 GT/s: Our analysis shows that a static selection does not work for all channels due to variations in the receiver design, channel, as well as PVT
- Solution: Need to adjust each transmitter at the by its corresponding receiver in a fine-grain fashion (coefficients)
  - Need to do it once and store it for use on every entry to 8GT/s
  - Must start with some predefined value set by platform characteristics
  - Dynamic adjustment after that
  - Our analysis shows this approach results in working silicon
Tx EQ coefficient Optimization vs. Pre-set example

- The eye diagram on the left was the result of using the best pre-set Tx EQ values.
- The eye diagram on the right was same channel with optimized Tx EQ coefficients.
- The green contour shows the BER eye at 1e-12.
- Eye width opening increased from 7ps to 16ps (over 50% more Eye Width)
  - Both assumed a Tx EQ step size resolution of 1/32
  - Channel: 2 connector topology 18” pin-pin
  - Both used same Rx EQ that was re-optimized for each case.

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Sample Transmitter and Receiver

Link Layer

Framing Logic

Lane Reversal/ Degradation

LFSR & OS Generator

Encoder

Mux

P2S, Driver

Considerations: 1 byte offset to Link Layer for TLPs (EDB). Seq # aligned to LCRC

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Summary & Call to Action

- Overview of Logical PHY based on Intel analysis, simulations, and experience
- 128b/130b Encoding definition
- Equalization mechanism needed
- 25% bandwidth advantage with new encoding over 8b/10b encoding with enhanced reliability
- Track the PCI Express* (PCIE*) 3.0 Spec development in the PCI-SIG and at www.pcisig.com
- Track the PCIE PIPE Spec development at www.intel.com/technology/pciexpress/devnet
  - Plan for products accordingly
Additional Sources of Information on This Topic

• Other Sessions / Chalk Talks / Labs:
  – **TCIQ002** Q&A: PCI Express* 3.0 Technology
  – **TCIS006** PCI Express* 3.0 Technology: Device Architecture optimizations on Intel Platforms
  – **TCIS008** Electrical requirements for designing PCIe* 3.0 ASICs on Intel platforms
  – **USBS002** USB 3.0 Architecture and PHY Interface (PIPE) Specification Updates

• Demo/Booths:
  – PCI Express* Technology Community

• Additional Web-based Info:
  – [www.pcisig.com](http://www.pcisig.com)
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